to conventional single-frequency sampling techniques. As noted above, a sensing zone may correspond to a single electrode 112, or a common modulation signal 110 may be provided to multiple electrodes 112 to create a larger sensing zone that effectively behaves as a single "electrode" for purposes of subsequent demodulation and computation. The modulated waveform 110, being the function of a distinct digital code 106, uniquely identifies the sensing zone to which it is applied, thereby allowing ready application of CDM and other conventional spread-spectrum techniques. FIG. 3 shows an exemplary spectral plot 300 that emphasizes the frequency domain differences between the spectrum 302 of carrier signal 111 and the spectrum 304 of modulated signal 110. In contrast to a single-frequency carrier signal 302, the multi-frequency spectrum 304 of modulated signal received on 114 is much wider. Because the spectrum 304 of the modulated signal distributes available power across a much wider sensitivity band, the effects of narrowband noise signals 306 at or near any particular frequency of interest are significantly reduced. That is, if a spurious noise signal 306 happened to occur near a singlefrequency (or narrowband) carrier signal 302, any electrical effects present within the sensing channel 113 could be overwhelmed by the noise. Moreover, adverse effects of wider-band noise 308 or interference from other modulated electrode channels 310 can be mitigated through spreadspectrum techniques, as shown in spectral plot 350 of received signal 116. Plot 350 shows that the demodulated signal 352 (corresponding to coupling of a channel 113 and/or the presence of object 121 near the demodulated sensing region) is contained within a relatively narrow frequency band, whereas signals 354 received from other channels are spread across a wider band. Both wideband noise 308 and narrow band noise 306 are similarly spread across wider frequency bands 356 and 358 in the demodulated signal. By increasing the bandwidth of the applied modulated signal 110, then, the signal-to-noise ratio in the demodulated signal 116 is improved dramatically. The demodulation in turn spreads the noise outside of the signal band, which then becomes quite narrow, thereby allowing the desired signal portion 352 to be readily extracted by a narrow band filter or the like.

[0041] This concept can be further exploited by selecting digital codes 106 that avoid known sources of noise. That is, digital codes 106 of any bit length may be applied to carrier signal 111 to create spectral "gaps" at frequencies known to be susceptible to spurious noise. By applying conventional Fourier analysis (e.g. using a simple fast Fourier transform (FFT) or the like), digital codes can be selected to create modulation signals 110 having many desired spectral characteristics. Codes applied to any modulated electrode 112 can be modified during operation (e.g. step 210 in FIG. 2) and/or may be pre-selected to avoid expected or observed noise in resultant signals 116. Alternatively, the particular codes 106 applied to one or more electrodes 112 may be randomly, pseudo-randomly, deterministically or otherwise modified during sensor operation, thereby statistically filtering any noise present within resultant signals 116 or demodulated signals 118 over time. Similarly, particular spatial frequencies or spatial positions can be emphasized (e.g. with code gain) or filtered out by the codes applied to different modulation electrodes. Code shifting during operation may provide other benefits (e.g. discrimination of or resilience to moisture) in addition to reducing the effects of noise or spurious effects of non-input objects (palm etc). In various equivalent embodiments, the frequency, phase, amplitude and/or waveform of carrier signal 111 may be adjusted in place of or in addition to modifying digital codes 106.

[0042] With reference now to FIG. 4, spread spectrum techniques that simultaneously apply multiple coded modulation signals 110A-D to various electrodes 112A-D are able to identify multiple presences 408, 410, 412 of objects 121 located within sensing region 101. Multiple objects may correspond to the presence of multiple fingers on a touchpad, for example, a palm resting on sensor 100 during use, simultaneous presence of a finger and a stylus, and/or any other combination of inputs. Electrical effects resulting from the various presences 408, 410 can be conceptually projected along one or more axes 404, 406 to identify the relative positions of the objects along that axis, as shown in FIG. 4. That is, peak value(s) of electrical effects can be correlated to relative positions of objects 121 with respect to sensing region 101. In the example of FIG. 4, a finger 408 may be identified by increases in electrical effects projected along an "X" axis 404 and a "Y" axis 406. By correlating the relative X and Y positions of peak electrical effects, the location of presence 408 can be correlated in two dimensions (or any other number of dimensions). Similarly, the example of FIG. 4A shows a larger area indicating a second presence 410 that results projections of electrical effects in axes 404 and 406. These multiple projections of electrical effect can be additionally correlated to identify images (e.g. "outlines") of objects 121 present within region 101. Taking this concept further, one or more images 408, 410 may be subsequently processed as appropriate. Presence of multiple fingers within region 101 may be used to perform scrolling, mode selection or other tasks, for example. Similarly, if an image can be identified as resulting from a user's palm (or another undesired portion of the user's body), that image 410 can be subsequently rejected in future processing, such as reporting of positional information or other output sig-

[0043] In the exemplary embodiment shown in FIG. 4, the two axes 404, 406 generally correspond to portions of modulated electrodes 112 or their associated channels shown arranged in two approximately orthogonal directions as in FIG. 1B. Alternate embodiments, however, may include any number of electrodes 112 arranged in any overlapping, non-overlapping, matrix or other arrangement. An example of a sensor 500 with overlapping electrodes 112A-G arranged in two dimensions is shown in FIG. 5. In such embodiments, electrical effects on received channels can effectively be independently measured at each crossing of the electrodes in two directions (e.g. X and Y directions corresponding to axes of 404, 406 in contour plot 400), with the results correlated in controller 102 to provide a twodimensional representation or image of object 121 rather than two one-dimensional "silhouettes" like 404 and 406. In such cases, electrodes arranged in the first direction (e.g. electrodes 112A-C) may be modulated at separate times from electrodes arranged in the second direction (e.g. electrodes 112D-G), with one or more independent received signals 116 at any one time from either set of electrodes (e.g. electrodes 112D&F) being provided to demodulator 117 via a multiplexer 502. FIG. 5 shows the various electrodes 112A-G coupled to both modulator 107 and demodulator 117 via a multiplexer 502. The multiplexer may also connect